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## **CPC (Cosmics Personality Card) for the L3+Cosmics experiment.**

The CPC replaces the MPC in L3, which is the interface of the muon chambers signals to the muon trigger. The CPC splits the signals from the discriminators in the muon system in L3. The old functionality of the MPC is copied into new devices and interfaces to the existing trigger system. Majority logic is added for a new trigger system for L3+Cosmics and TDCs measure the relative timing of muons. A stand alone read-out system in VME is used to read the data independent from the existing L3 data acquisition system.

This paper describes CPC version 3.

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## Introduction

For the L3+Cosmics experiment the existing L3 muon chambers are used. The signals from the L3 discriminators are used to generate a trigger and to measure their relative timing. A convenient place was found at the back of the existing muon TDC crates. There the MPCs (Muon Personality Cards) connect the FastBus TDCs to the first and second level muon trigger. The signals on the auxiliary connector are a copy of the discriminator signals that come in at the front panel of the TDC cards, converted to TTL signals.

The MPCs are replaced by CPCs (Cosmic Personality Cards). On these cards the old (MPC) functionality has been redesigned to occupy less space and consume less power. The added functions are the majority logic needed for the L3+Cosmics trigger and a TDC system for time measurements independently from the existing L3 system.

First a prototype (CPC0) was installed to demonstrate that the wire signals can be split to provide the L3+Cosmics system with timing signals from the muon chambers, without disturbing or altering the L3 muon trigger interface. These are now replaced by the final CPC.

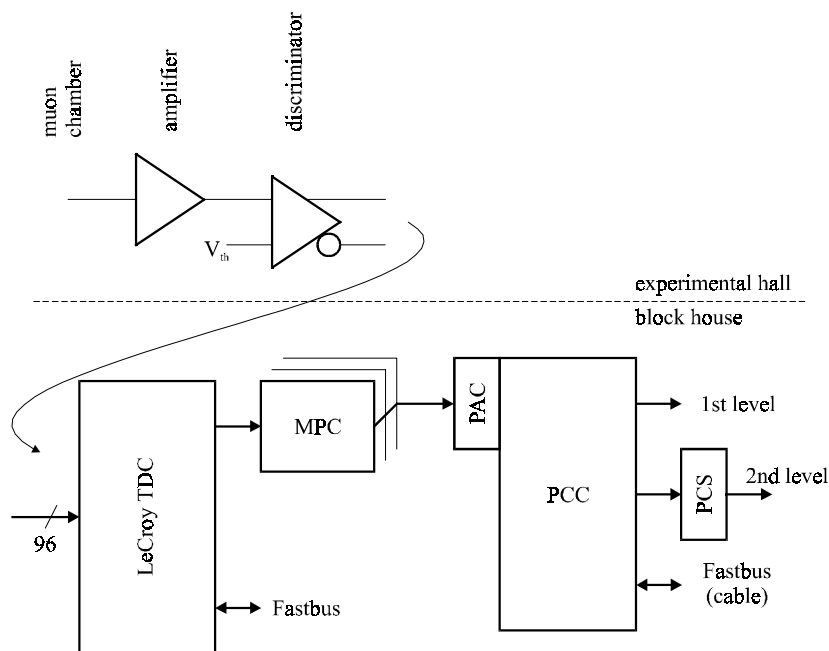


Figure 1: Signal flow from muon chambers

### About V3, modified V2

On version 3 (or modified version 2) the quality of the 40 MHz clock, driving the TDCs, is further improved. The buffer that used to be located on a small modification card, mounted on the CPC\_V2, is replaced by a RoboClock (Cypress CY7B991). On the modified version 2 this is again a small modification board. On version 3, it is integrated on the CPC itself.

Via a register in the majority logic (IC8) it can be checked whether the 40 MHz clock is running (see Table 9). There is a difference here between the modified version 2 and version 3. In version 2 the incoming clock from the *front end link* is monitored. In version 3 the clock from the RoboClock is monitored. It is possible that the RoboClock still runs at its minimum frequency when the Front End link is absent. This can only be checked by observing the *lock error* of the TDC or checking the communication via the *.front end link*. If the CPC powers up without the input clock, the RoboClock is quiet. When the clock is removed during running, its frequency drops to the minimum lock frequency, which is somewhere between 10 and 20 MHz (guaranteed minimum frequency is 20 MHz).

New test pulse functions have been implemented. This is in V3, as well in the modified V2. The external testpulse, running on the JTAG cable, is now also available on the Read Out Controller (IC7). And there is a connection, running from the ROC (Read Out Controller) to the Control-Mach (IC3). This makes it possible to generate test pulses to the TDCs and majority inputs from the ROC. From the external testpulse, the ROC can generate a trigger to the TDCs with a fixed delay in respect of the TDC's input pulses. The feature is added for testing purposes.

This test pulse must be enabled in the Control-Mach (see Table 5, control register). Though the definition of the control register has changed, it is compatible with older routines, choosing between the previously existing test pulse sources.

Modified V2 cards can easily be recognized by the copy of the serial number on the back side of the RJ45 connector and the yellow *connect* LED, which used to be red. The serial numbers of V2 range from x011 until x04c and the serial numbers for V3 start at x051.

## MPC functionality

The MPC is described in the paper “The L3 muon trigger interface”<sup>[1]</sup>. Since the functionality is extended in the case of the CPC, it is described here briefly.

The MPC functions can be reached via the LeCroy FastBus TDC. Nine secondary addresses are used to reach the data registers and status/ control address. For the CPC a new address is implemented (C,3). Reading this location will return the card’s fixed address used in the addressable JTAG chain (10 bits). This is the same as the card’s serial number. The identification register can be used to store a 6 bit position identification via FastBus into the CPC. This will be further described in the JTAG description.

data registers		
Address	Read	Write
C,2	Status	Control
C,3	Card address	Identification
C,8...C,B	MPC registers	Add pattern to MPC registers
C,C...C,F	Ored inputs MPC	Replace pattern MPC registers

(C,2 is shorthand for address C0000002)

Table 1: MPC registers

A flip-flop on the MPC determines whether the card is controlled by the PCC (Personality Card Controller) or from FastBus through the LeCroy TDC. The PCC overrides operations from FastBus. When the PCC is master, the data registers cannot be accessed and an SS6 response is returned. The 48 bit data register is split in 4 parts of 12 bits, due to limitations on the auxiliary FastBus connector.

data	control (write C,2)
4	Enable collect
5	Disable collect
8	Set all registers
9	Reset all registers
C	TDC becomes master
D	let PCC be master

Table 2: MPC Control register

Bit	Status (read C,2)
0	PCC is reading the MPC
1	PCC is master of MPC
2	Collect enabled
3	Power failure on daisy chain circuit

Table 3: MPC status register

The enable/ disable collect commands set and reset a flip flop on the MPC. When this flip flop is set, the 48 registers collect and store pulses from the inputs in the same way as the registers are used by the PCC to collect wire signals. When the PCC becomes master, the flip flop is reset automatically. Only when the TDC is master of the MPC, the data registers can be accessed and the collect flip flop can be controlled. The TDC cannot become master of the MPC when the PCC accesses the MPC at that moment.

<sup>1</sup> The muon trigger interface for L3: MPC, PAC, PCC-P, PCC-Z, PCS and PCT, september 1989, H. Groenstege NIKHEF-ET.

## CPC

### Signal Split

The signal split to the added functionality (majority logic and new TDCs<sup>2</sup>) is implemented in the same devices that contain the MPC functionality. This strongly reduces the pin count on the board. The old functionality has not been changed. The MACH 466 devices also contain two sets of registers. One set of 96 bits is used to disable noisy channels (only for the new functionality) . The other set of 96 bits is used to inject test patterns upon reception of a test pulse.

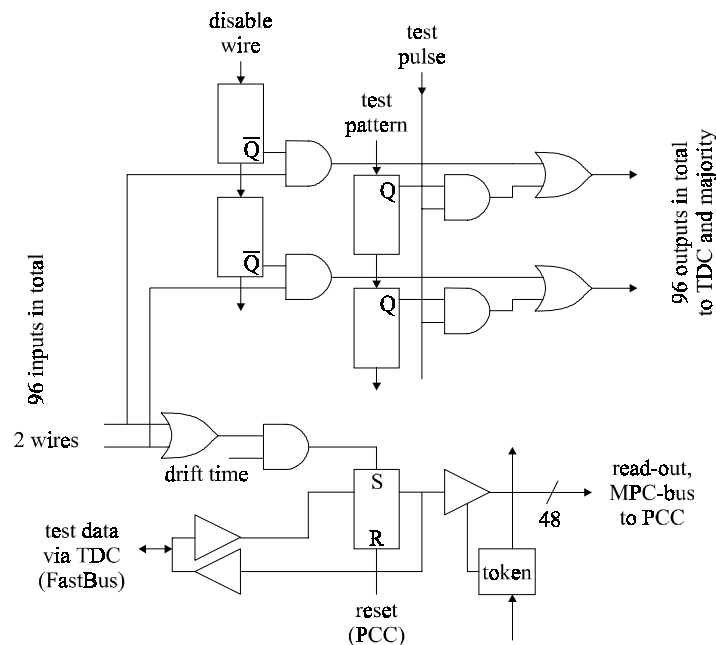


Figure 2: CPC signal split, mask and test pulse

The two set of registers are configured to behave as a JTAG port, which is controlled by a MACH 231 device. The bits are shifted with LSB (Least Significant Bit) first. Upon power-up, the registers are initialized automatically. All wire signals are enabled to the output and the test pattern is reset. However, this cannot be guaranteed, since it requires the power supply to rise monotonously. A power failure can be detected by checking a bit in the Majority Logic, see page 14 (Majority Logic, Control).

### Majority logic

IC 8 contains the majority logic. It can run in two modes:

- The 96 inputs are compared with a single threshold.
- The 96 inputs are divided in an upper and a lower half (48 inputs each).

An input signal, longer then app. 15 ns, is recorded. All signals are stretched to app. 1.2  $\mu$ s. Every 200 ns the total number of hits is determined. This number is compared to the programmable threshold and may thus generate a hit. This hit signal belongs to one of the

<sup>2</sup> A 32 channel general purpose Time to Digital Converter, J. Christiansen CERN/ECP-MIC.

three muon chamber layers; MI, MM or MO. A register setting determines which output has to be activated. This output is a differential implementation of an open collector output. In this way all CPCs in a crate can pull this line. This line then runs to the CTT (Cosmic Trigger and Timing) module, where the Cosmic trigger is generated.

In every P-crate there is one location where signals from two different muon chamber layers are connected to one FastBus TDC. In that case the majority logic works in the splitted mode, with two times 48 channels. Two output signals are generated in the same way as described above and can be routed (software controlled) to the MI, MM and MO outputs.

The majority logic chip (IC 8) receives the 40 MHz clock from the read out chip (IC 7). This clock is divided internally to create the clock for the pulse stretchers and other pipe line registers. This divider can be programmed to divide from 7 to 10.5 in eight steps. The setting is done via JTAG. The pipe line clock periods thus vary from 175 to 263 ns, in 12.5 ns steps. The pulse stretchers are active for seven of these cycles. So the one shots run for 1050 to 1575 ns, plus one clock period. The input signal is first stored in a register. On the next clock cycle the one shot starts. This covers the maximum drift time plus one clock cycle for the synchronization uncertainty. Also one clock cycle overlap is needed to determine the total of active one shots. The drift time range is then 1050 to 1575 ns. For information how to set the stretchers see page 13. The reset signal from the FELink can be used to synchronize the clock dividers on the different CPCs.

For the scintillators the chip that normally contains the majority logic (IC 8) has an other contents. It logically ORs every eight inputs, which results in 12 signals. These signals can be used by three outputs to form larger ORs. These three outputs are first stretched and then delayed. The stretcher is a retriggerable one-shot that is fired by the rising edge of the OR signal. The delay is made by means of a shift register. For the settings see Table 14: Scintillator, stretch and delay.

### ***Read out control logic***

The CPC Read Out Controller (ROC) in IC7 receives the clock (40 MHz) and control signal (Reset/Trigger) from the NIMROD. The control signal encodes the Reset and Trigger signals using a 3 bit pulse width scheme. Holding the control line high for one clock cycle, followed by a low level for at least two clock cycles encodes the Trigger signal. Holding the control line at high level for three consecutive cycles encodes the Reset signal.

The Reset and Trigger signals are distributed to the TDC chips and the Majority logic chip. The Reset signal clears all internal counters and the data memory in the TDC chips. The Trigger signal is used by the ROC to initiate the read out of the TDC chips. The data from the TDC's is read out using a token ring scheme and the data is formatted by adding some identification bits, serialized and sent to the NIMROD via the Front End Link. Data is transferred via two pairs of Low Voltage Differential Signal (LVDS) drivers using the Data-Strobe (DS) protocol. The bit rate is 40 Mbit/s and each data word is composed of a start bit, 32 data bits, a parity bit and one stop bit. This results in a maximum data rate of approximately 1 Mword/s.

Provisions have been made to be able to read out the Majority Logic as another node in the token ring. This would allow the Majority Logic to send one word of information into the data stream after a Trigger, whenever that becomes useful.

**Card identification**

The card address (10 bits) is defined when it is tested after production. This is fixed. The address is the same as its serial number written on the white space. The address is used by the addressable JTAG controller to uniquely address a single card. When two cards in the FastBus crate are swapped, this cannot be detected via the addressable JTAG branch. Two provisions are implemented to deal with the problem:

- Using a geographical read cycle in FastBus, the card's serial number can be read.

This serial number is used as an address in addressable JTAG operations. See the MPC register description on page 5 (Table 1: MPC registers).

- Using a geographical write cycle in FastBus, an identifier can be written into C,3.

It is useful to write the card location inside the FastBus crate into this register. The stored value can be read back via the addressable JTAG chain. See the FastBus control register description on page 5 and the JTAG register description on page 11 (Table 5: Mask/ test pattern, Instruction register).

Both methods require access to the L3 FastBus TDCs.

## JTAG

The CPC contains three JTAG ports. These are used for programming some of the devices, testing the printed circuit board and for the settings needed at run time. Some JTAG ports are the standard JTAG ports of the devices, others are created by programming the devices.

The pins on the dual row connectors are wired according to the MACH-PRO convention. The Vcc pin is connected to the 5 V supply of the CPC, via a 100  $\Omega$  resistor (4.7  $\Omega$  in case of the BST port). The TRST (Test bus Reset) is only used on the settings port. The optional program enable pins on the MACH devices are not used.

Pin	Signal	Pin	signal
1	TCK	6	Vcc
2	NC	7	TDO
3	TMS	8	GND
4	GND	9	TRST
5	TDI	10	NC

Table 4: MachPro connector

Port *MACH\_PROG* is used to program the MACH devices that emulate the MPC functionality and the signal split. This program is stored in EEPROM inside the devices. It is non-volatile, but it can be altered. IC 1 and IC 2 (MACH 466 devices) contain the same program. IC 3 is the controller of ICs 1 and 2 for the old MPC functionality, for the emulated JTAG chain and for several other functions. To program the devices, the MACH-PRO software (version 1.40f or later) is used.

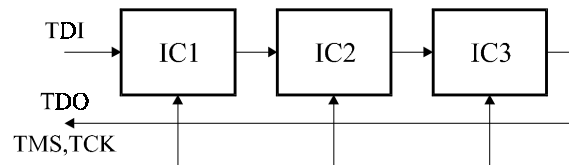


Figure 3: MACH\_PROG, programming port

The programs stored in IC1, 2 and 3 together create a second JTAG port on IC3, which is used to load the disable and test pattern registers, for example. These registers are shifted with LSB first.

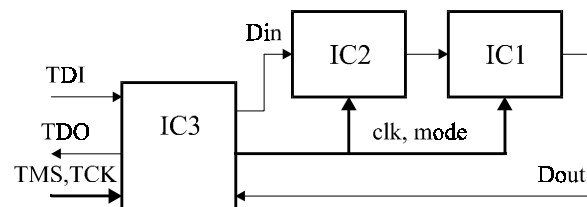


Figure 4: JTAG2, emulated port

### JTAG chains

The mask and test pattern registers and the majority logic and read-out chips contain emulated JTAG ports. Together with the TDCs they are connected to the addressable JTAG port (*Ajtag*) that will be used to do the various settings on the card. The addressable JTAG ports of all CPCs in one FastBus crate are connected to a JTAG controller, located in a VME crate. For testing, the addressable part can be bypassed, by connecting a JTAG cable to the port *Settings*.

Below the JTAG chain order and connector names on the CPC are given.

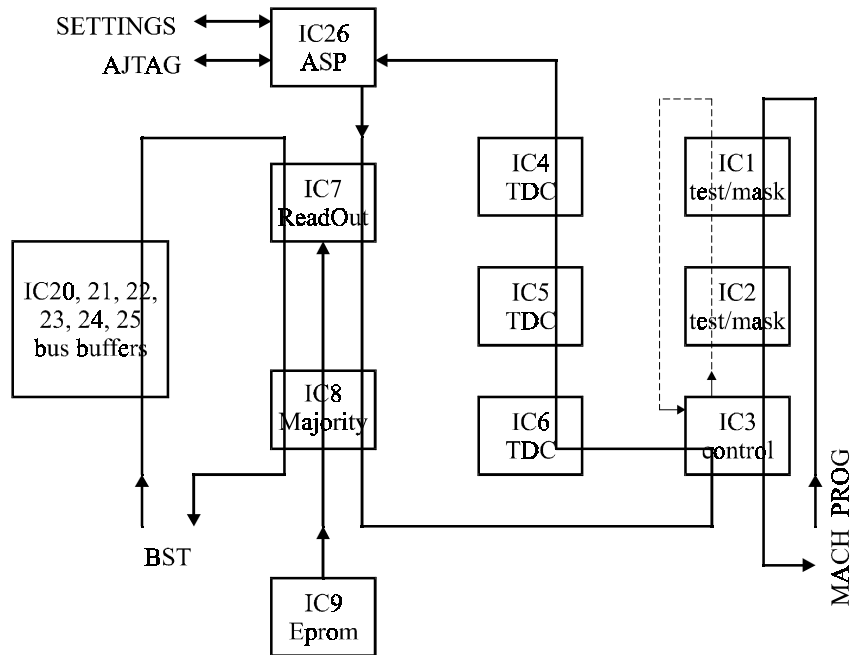


Figure 5: Chain orders and connector names

The BST (Boundary Scan Test) chain can be used to test the card electrically. In this case a test set-up exists for the MPC functionality. Nearly all other functionality can be tested via the *Settings* chain.

The EProm holds the files used to define the functionality of the Altera 10k20 chips, IC 7 and IC 8. These files are automatically loaded at power-up. The EProm is a OTP (One Time Programmable) version. The order of the programming files is first the majority logic and second the Read out controller functionality. The BST chain can be used to reprogram the Altera devices. The programming routine of the Altera software can be used without the hardware protection key. It requires a *byte blaster* download cable, connected to the parallel port of PC. Because the JTAG chain runs in the other direction, compared to the *passive serial load* chain (from the EProm), a valid program must reside in EProm to be able to use the *byte blaster*.

### Mask and test pattern

The combined functionality of ICs 1, 2 and 3, besides creating the MPC functionality, is to split the incoming signals to the TDCs and majority logic. They contain mask and test registers. These are loaded via JTAG.

IR	Instruction	Bits	R/W	Default
0000	Test pattern	96	R/W	0...0
0001	Mask register	96	R/W	0...0
0010	Control register	4	R/W	0000
0011	FastBus ID register	6	R	00 0000
Others	Bypass	1	R	0

Table 5: Mask/ test pattern, Instruction register

#### Test pattern (IR=0000)

After loading the test pattern, test pulses can be generated in the *Pause\_DR* state or the *Run-Test/Idle* state of the TAP controller. Efficient tests (a walking bit, for example) can be performed, by turning from *Pause\_DR* back to *Shift\_Dr*, instead of loading a complete new 96 bit pattern. This is possible, since the test pattern register does not have a shadow register. One does not have to pass through the *Update\_DR* state to activate the pattern. Usually the *Run Test/Idle* state is used to generate test pulses. For both states TMS is kept zero to remain in the state. If the state is just passed, to go to another state (respectively *Exit2\_DR* and *Select\_DR\_Scan*), no test pulses are issued. The source and the length of the pulses are determined by the contents of the control register.

#### Mask pattern (IR=0001)

Via this register noisy or unused channels can be disabled. Loading a one means disable this channel for the TDCs and the majority logic.

#### Control register (IR=0010)

The control register contains four bits.

Bit 2...0	Test pulse	Note
000	No test pulse	*1
X01	Test pulse from TCK	*2
X10	Test pulse from EXT (JTAG)	*2
X11	Test pulse from ROC	*2
100	Test pulse from instruction	*3

Table 6 : Testmode select

\*1: Power up state.

\*2: The TAP controller must be in state *Run-Test/Idle* for test pulse generation.

\*3: Writing X100 in the control register will generate a test pulse in the TAP controller state *Update\_DR*. Bit 2 is automatically reset in the next state of the TAP controller.

**Bit 3:** Disable test pulse cut. When set, the on board test pulses, initiated by one of the test pulse sources, have the same length as the input pulses. At power up, the bit is reset. In that case the test pulse width is limited to app. 125 ns.

**FastBus ID (IR=0011)**

The FastBus ID register is a 6 bit read only register. It is used to verify the geographical location of the CPC, see page 8 for the description.

**IR='others'**

When loading an other pattern then those described here, the function is *Bypass*. This is mandatory for IR=1111. The Bypass Register is one bit long.

**Majority logic**

IC 8 contains the majority logic. It is set up via JTAG. It can run in two modes:

- The 96 inputs are compared with a single threshold.
- The 96 inputs are divided in an upper and a lower half (48 inputs each). Each half has its own threshold register.

The instruction register is four bits long.

IR	Register	R/W	PU
0000	Threshold 0	R/W	0111 1111
0001	Threshold 1	R/W	0011 1111
0010	Set_up	R/W	0000 1111
0011	Control/ Status	R/W	0000 1X00
0100	Version	R	1000 0010
0101	Stretcher	R/W	0000 0011
Others	Bypass	R	0
PU describes the power up situation. This can also be forced by the reset bit in the control register.			

Table 7: Majority logic, Instruction register

All registers are 8 bits long, except for the bypass register, which is one bit long.

**Threshold register 0 (IR=0000)**

Threshold register 0 contains the threshold (7 bits) for the lower half of the inputs or the full 96 inputs. It is a read/ write register. During a write, bit 7 is ignored. When read back, bit 7 is 0.

**Threshold register 1 (IR=0001)**

Threshold register 1 contains the threshold (6 bits) for the upper half of the inputs. It is a read/ write register. During a write, bits 6 and 7 are ignored. When read back, bits 6 and 7 are 0.

**Set up register (IR=0010)**

The set up register determines which outputs are driven when the total number of hits exceeds the threshold. Bit 4 determines whether there are one or two outputs. At power up the bit is reset. In that case there is one output. Bits 1 and 0 determine lower half (or total) output selection. Bits 3 and 2 determine the upper half output selection. When the bit pairs are 11, no output is selected.

Bits	PU	Value	Result	Comment
1:0	11	00	Hit0 to MI	Hit0 is the result from the total or the lower half.
		01	Hit0 to MM	
		10	Hit0 to MO	
		11	No output	
3:2	11	00	Hit1 to MI	Hit1 is the result from the upper half.
		01	Hit1 to MM	
		10	Hit1 to MO	
		11	No output	
4	0	0	96 inputs	See also page 6.
		1	2*48 inputs	
7:5	000			Not used
PU describes the power up situation. This can also be forced by the reset bit in the control register.				

Table 8: Majority logic, Set up

The settings can be read back. Bits 7, 6 and 5 are zero. When the bit 4 is one and for both halves the same output is selected, the two hit results are logically ORed.

**Control/ Status register (IR=0011)**

Bit	PU	Description
0	0	Hit 0 Register
1	0	Hit 1 Register
2	X	Clock running
3	1	Power up reset monitor
4	0	(Differential mode)
5	0	Enable drivers
6	1	Enable synchronization
7	0	Simulate power up
PU: Power Up situation		

Table 9: Majority logic, Control

The hit 0 register (**bit 0**) in the control/ status register monitors whether the output of lower half (or the total number of channels) of the majority logic has seen a hit. A hit means that the number of active channels within the maximum drift time period has exceeded the programmed threshold. The bit is reset when read (destructive read). **Bit 1** does the same for the upper half.

**Bit 2** monitors the 40 MHz clock. It is reset when the TAP controller passes the *Run-test/Idle* state. It is set by clocking a one by means of the 40 MHz clock, during the time the TAP controller moves to *Capture\_DR*. There is a difference here between the modified version 2 and version 3. In version 2 the incoming clock from the Front End link is monitored. In version 3 the clock from the RoboClock is monitored. It is possible that the RoboClock still runs at its minimum frequency when the Front End link is absent. This can only be checked by observing the *lock error* of the TDC or checking the communication via the *.front end link*. If the CPC powers up without the input clock, the RoboClock is quiet. When the clock is removed during running, its frequency drops to the minimum lock frequency, which is somewhere between 10 and 20 MHz (guaranteed minimum frequency is 20 MHz).

**Bit 3** registers a (hardware) power up reset. This signal is generated by a voltage monitor chip connected to the supply voltage of the board. The bit is reset when read (destructive read)

**Bit 4** is not used.

**Bit 5** is the output enable of the three majority output drivers.

**Bit 6** enables synchronization of the one-shots in the majority logic. The reset signal from the FELink is passed to the majority logic, via the read out chip. The front edge of the reset signal is used to synchronize the pipe line clock. If the reset signal is longer then two clock periods (of 25 ns), the majority output drivers are disabled until the reset drops again.

Register **bit 7** is used to simulate a power up reset. It must be set to one and than reset again. It does not affect the TAP controller. The other bits in the register and other registers are set to the power up situation.

#### **Version register (IR=0100)**

The Version register is used to identify the version of the programmed functionality. It is a read only register. For the version described here it reads 1000 0010.

### Stretcher register (IR=0101)

The Stretcher register determines the speed of the one shots. Since the input pulses have to be synchronized with the digital one\_shot's clock, the total length must be the drift time plus one clock cycle. The clock for the one\_shots is derived from the 40 MHz clock, coming from the NIMROD. This clock is divided to obtain the pipe line clock. Six clock cycles from the divider are used for the drift time by the one shot and one more for the synchronization uncertainty. For the L3 muon chambers the drift time is app. 1.2  $\mu$ s.

Bit 2:0	+ N	Pipe Clock [ns]	Drift Time [ns]	Stretch Time [ns]
000	7	175	1050	1225
001	7.5	188	1125	1313
010	8	200	1200	1400
011 *	8.5	213	1275	1488
100	9	225	1350	1575
101	9.5	238	1425	1663
110	10	250	1500	1750
111	10.5	263	1575	1838

\*: Power up setting.

Table 10: Majority logic, input stretcher timing

The halves in the table are created by alternating dividing by N and N+1. The stretch time is the maximum drift time plus the input synchronization time. This is the time that the one-shot runs.

**Bit 3** disables the time over threshold output. Comparators in the majority logic check whether the amount over one-shots running is more than specified in the threshold register. This is done every pipe-line clock. Thus the output of the comparator is time over threshold. The rising edge of this signal starts a one-shot. The one-shot then runs for a period specified in bits 7 to 4. When a new rising edge of the time over threshold is detected while the one-shot is already running, the one-shot is restarted. So it is a retriggerable one-shot. Restarting does not introduce gaps in the output signal. The output of this one-shot is logically ORred with the time over threshold signal and forms the majority output towards the CTT. Setting bit 3 disables the ORing of the time over threshold. In that case only the one-shot output is seen at the output. At power up, bit 3 is reset.

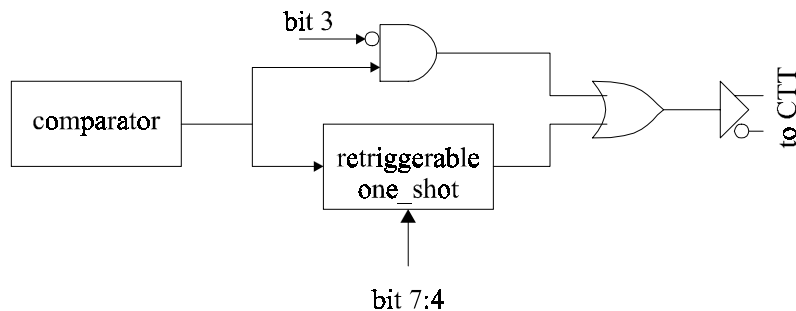


Figure 6: Majority output logic

ns	bits 2:0 (Pipeline clock, see Table 10)							
Bits 7:4	000	001	010	011*	100	101	110	111
	175	188	200	213	225	238	250	263
0000	175	188	200	213	225	238	250	263
0001	175	188	200	213	225	238	250	263
0010	350	376	400	426	450	476	500	526
0011	525	564	600	639	675	714	750	789
0100	700	752	800	852	900	952	1000	1052
0101	875	940	1000	1065	1125	1190	1250	1315
0110	1050	1128	1200	1278	1350	1428	1500	1578
0111*	1225	1316	1400	1491*	1575	1666	1750	1841
1000	1400	1504	1600	1704	1800	1904	2000	2104
1001	1575	1692	1800	1917	2025	2142	2250	2367
1010	1750	1880	2000	2130	2250	2380	2500	2630
1011	1925	2068	2200	2343	2475	2618	2750	2893
1100	2100	2256	2400	2556	2700	2856	3000	3156
1101	2275	2444	2600	2769	2925	3094	3250	3419
1110	2450	2632	2800	2982	3150	3332	3500	3682
1111	2625	2820	3000	3195	3375	3570	3750	3945
*: Power up setting								
*: Power up setting								

The minimum pulse width of the majority output signals is one pipeline clock cycle.

Table 11: Majority output stretcher

#### IR='others'

When loading an other pattern in the Instruction Register then those described here, the function is *Bypass*. This mandatory for IR=1111. The Bypass Register is one bit long.

**Scintillator CPC**

The programming of IC 8 is different for the scintillator CPC, used in the *auxiliary box*. The chip that normally contains the majority logic (IC 8) has an other contents. It logically ORs every eight inputs, which results in 12 signals. These signals can be used by three outputs to form larger ORs. These three outputs are first stretched and then delayed. The stretcher is a retriggerable one-shot that is fired by the rising edge of the OR signal. The delay is made by means of a shift register.

Also IC 1, 2 and 3 behave different. The FastBus interface is removed and the FB\_ID will always return 00 0000 via Jtag. The 96 inputs are positive inputs for this version, instead of active low as needed for the interface to the LeCroy TDC. The mask and test pattern act the same as in the normal CPC.

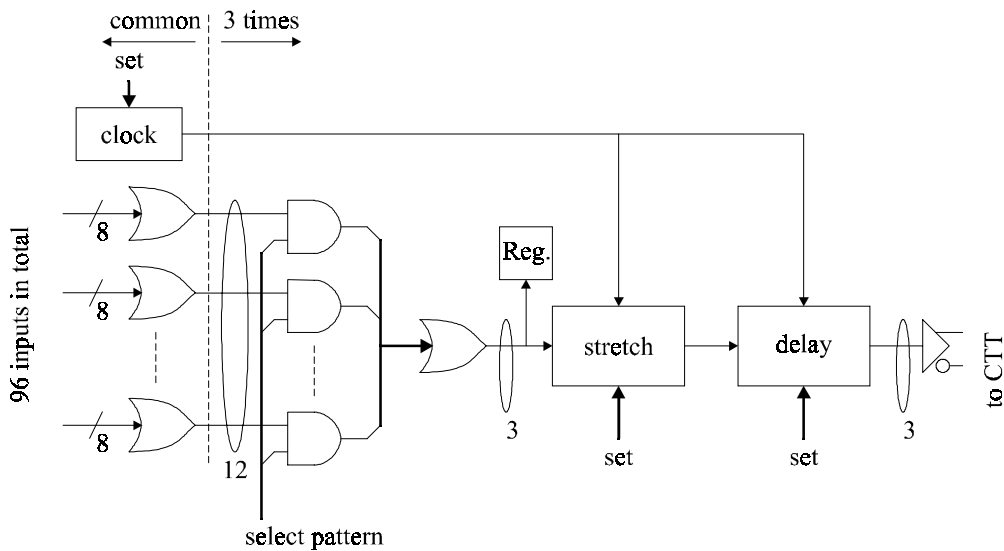


Figure 7: IC 8 for scintillator CPC

IR	Register	R/W	PU
0000	Select pattern 1	R/W	0000 0000 0000
0001	Select pattern 2	R/W	0000 0000 0000
0010	Select pattern 3	R/W	0000 0000 0000
0011	Control/ Status	R/W	0100 0111 X000
0100	Version register	R	1100 0000 0010
0101	Clock divider	R/W	0000 0000 0011
0110	Stretch register	R/W	1001 1001 1001
0111	Delay register	R/W	0110 0110 0110
1000	Test point register	R/W	0000 0111 1111

PU describes the power up situation. This can also be forced by the reset bit in the control register.

Table 12: Scintillator, Instruction register

**Select pattern (IR=0000, 0001, 0010)**

Every eight inputs are logically ORred. For the three outputs, these patterns determine which of these twelve groups are used to create the complete OR. The select pattern is numbered 1 to 3, according to the octant numbers. On the board and in the schematics, outputs 0, 1 and 2 are called MI, MM and MO respectively.

**Control/ status register (IR=0011)**

The control/ status register

Bit	PU	Description
0	0	Hit register 1
1	0	Hit register 2
2	0	Hit register 3
3	X	Clock running
4	1	Differential mode
5	1	Enable output drivers
6	1	Enable synchronization
7	0	NoOR
8,9	00	Not used
10	1	Power up reset monitor
11	0	Simulate power up
PU: Power Up situation		

Table 13: Scintillator control register

**Bit 0, 1 and 2** register whether one of the outputs has seen a hit. The bits are reset when read (destructive read).

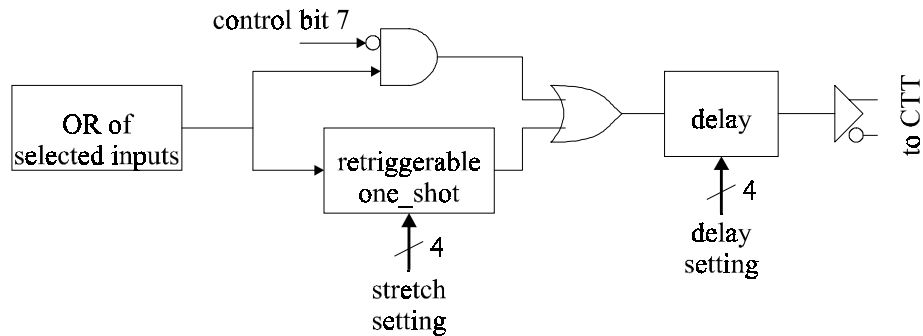
**Bit 3** monitors the 40 MHz clock. It is reset when the TAP controller passes the *Run-test/Idle* state. It is set by clocking a one by means of the 40 MHz clock, during the time the TAP controller moves to *Capture\_DR*. There is a difference here between the modified version 2 and version 3. In version 2 the incoming clock from the Front End link is monitored. In version 3 the clock from the RoboClock is monitored. It is possible that the RoboClock still runs at its minimum frequency when the Front End link is absent. This can only be checked by observing the *lock error* of the TDC or checking the communication via the *. front end link*. If the CPC powers up without the input clock, the RoboClock is quiet. When the clock is removed during running, its frequency drops to the minimum lock frequency, which is somewhere between 10 and 20 MHz (guaranteed minimum frequency is 20 MHz).

**Bit 4** controls the mode of operation of the output drivers. The outputs are truly differential or the differential implementation of an open collector output. For the scintillator CPC it should be set (power up, default situation).

**Bit 5** is the output enable of the three output drivers. For the scintillator CPC the bit is set during power up.

**Bit 6** enables synchronization of the clock divider. The reset signal from the FELink is passed to the stretcher logic, via the read out chip. The front edge of the reset signal is used to synchronize the pipe line clock. If the reset signal is longer then two clock periods (of 25 ns), the output drivers are disabled until the reset drops again.

**Bit 7** disables the logical ORing of the original input signal with the output of the one-shot.



**Bits 8 and 9** are not used.

**Bit 10** registers a (hardware) power up reset. This signal is generated by a voltage monitor chip connected to the supply voltage of the board. The bit is reset when read (destructive read).

**Bit 11** is used to simulate a power up reset. It must be set to one and then reset again. It does not affect the TAP controller. The other bits in the register and the other registers are set to the power up situation.

#### **Version register (IR=0100)**

The version register is used to identify the version of the programmed functionality. It is a read only register. For the version described here it reads 1100 0000 0010. Version 1100 0000 0001 does not contain the test point selector.

#### **Clock speed (IR=0101)**

The pipe line clock speed is determined by three bits. It works the same as for the majority logic. For the definition see Table 10: Majority logic, input stretcher timing.

#### **Stretch register (IR=0110)**

The output stretch register determines the length of the output pulse in pipe line clock cycle units. The minimum is one clock cycle. The maximum is 15 clock cycles. The values for the three outputs can be set to different values. See the table below.

**Delay register (IR=0111)**

The delay register determines the length of the output pulse in pipe line clock cycle units. The minimum is one clock cycle. The maximum is 15 clock cycles. One more clock cycle is used in the pipe line structure. The values for the three outputs can be set to different values.

		Bit groups for stretch and delay registers		
Output		3	2	1
Bits		11:8	7:4	3:0
default	Stretch	1001	1001	1001
	Delay	0110	0110	0110

Table 14: Scintillator, stretch and delay

Use Table 11 to determine the values in ns.

**Test point register (IR=1000)**

The test point register selects one of the 96 inputs to be routed to pin 7 of connector Diag8 on the CPC.

Test point register		
Write	Read	Power up
XXXX XSSS SSSS	0000 OSSS SSSS	0000 0111 1111

Table 15: Scintillator, test point selector

Valid range for input selection is 0 to 95. If a higher value is written (96...127), the output is zero. Please note that the input is taken after the mask and test pattern section of the CPC.

**Diagnostics connector**

Diagnostics connector Diag 8 carries the following signals:

Pin	Name	Use
1	Vcc	+5 V
2	Clk_div	Divided clock used by IC 8
3	Dia8p9	In CPC V2 used to connect to IC 7
4, 5, 6	TPM4, 3, 2	Not used
7	TPM1	Test point output, see <i>test point register</i>
8	_PUR	Power up reset monitor of IC 8 only
9...12	TPM7, 8, 5, 6	Not used
13	_A_Hit	Stretched hit signal
14, 15, 16	MO_D, MM_D, MI_D	Logic one / active in scintillator CPC
17, 18, 19	MO_E, MM_E, MI_E	Majority outputs / logic one in scintillator CPC
20	Gnd	0 V

### **Read out control logic**

The chip IC 7 contains the Read Out Control logic (ROC). It is set up via JTAG and the clock, trigger and reset signals are handled to acquire and send out the TDC data via the Front End Link.

### **Front End Link**

The clock, trigger and reset signals are sent from the Cosmic Trigger and Timing unit to the CPC cards via the Front End Link cable (standard FTP-type cable with RJ45 connectors) using differential LVDS drivers and receivers. The clock is transmitted via one pair of wires, the other signals are time-encoded over one other pair of wires. The encoder uses three consecutive cycles of the 40 MHz clock and uniquely defines the four signals as follows:

<b>t1</b>	<b>t2</b>	<b>t3</b>	<b>Description</b>
0	0	0	no operation
1	0	0	Trigger
1	1	0	bunch count reset: resets only the TDCs
1	0	1	event count reset: reset the ROC and the TDCs
1	1	1	global reset: resets the ROC and the TDCs

### **Data format**

The data format that the ROC generates is described in a separate document called “The L3 + Cosmics Data Format” by T. Wijnen, B. Petersen and C. Timmermans.

### **JTAG**

The JTAG instruction register is four bits long and it is used to select one of the internal data registers.

<b>IR</b>	<b>Register</b>	<b>Mode</b>	<b>Default</b>
0000	User identity	R/W	1111 1111
0001	Control	R/W	1100 0000
0010	Setup	R/W	0010 0000
0011	Version	R	1000 0110
0100	Diagnostics	R	0000 0000
0101	DAC	R/W	1000 0000
0110	Mode	R/W	0000 0000
0111	Rand	R/W	0000 0000
1111	Bypass	R	0
Others	Bypass	R	0

Table 16: Read Out Control, JTAG register selection

All data registers are 8 bits long, except for the Bypass Register, which is one bit long. The default value is the contents of the data registers at power up.

#### **User Identity Register (IR=0000)**

The eight bits User Identity register is used to identify the card when read out via the *front end link* (FELink). It can be filled with e.g. the crate number and card position.

In the current version (dated 12-Mar-1999) this information can not be read out via the FELink!

#### **Control Register (IR=0001)**

The Control register is used to control the functions of the ROC. The contents of the register may only be changed when the trigger system is disabled, because otherwise the ROC functions could be altered or aborted while it is processing data. The individual bits have the following meaning:

<b>Bit</b>	<b>Function</b>	<b>Default</b>	<b>Description</b>
0	CR0	0	Clock speed select for ROC test pulse
1	CR1	0	Clock speed select for ROC test pulse
2	EnTTrigger	0	Enable ROC test pulse trigger
3	EnMTrigger	0	Enable "internal" Majority Trigger signal
4	EnETrigger	0	Enable external Trigger and Reset signal from test connector
5	EnART	0	Enable the artificial Reset/Trigger signals ( these are generated via EnTTrigger, EnETrigger and EnMTrigger )
6	Inhibit FE Link	1	Disable Reset/Trigger signals via FELink
7	Inhibit ROC	1	Disable ROC (holds the Reset signal for the ROC and TDCs

Table 17: Read Out Control: Control register

After power up the Control register is preset to "1100 0000" and it is thus holding the Reset signal for the ROC chip and the three TDC chips in a steady active state.

**Setup Register (IR=0010)**

The Setup register is used to setup the different modes and options in the ROC. The contents of this register may be altered only when the trigger system is paused, because any change causes the token ring read out scheme to be altered immediately. The individual bits have the following meaning:

Bit	Function	Default	Description
0	EnTDC	1	Enable the read out of TDCs
1	Unused	0	
2	Unused	0	
3	EnMAJ *	0	Enable the read out of the Majority Logic
4	Unused	0	
5	EnEOG	1	Enable the End Of Group word
6	EnFake	0	Enable the fake TDC words
7	EnEcnt	0	Enable the Event Count word

Table 18: Read Out Control: Setup register

The function bits marked with "\*" are not implemented in this version of the ROC chip, dated 12-Mar-1999.

**Version Register (IR=0011)**

The Version register is used to identify the version number of the programmed functionality in this chip. It is used to determine which "firmware" is present on this CPC card. This version (12-Mar-1999) of the ROC will identify itself with a value of "1000 0110". At power up this value is loaded from the configuration PROM.

**Diagnostics Register (IR=0100)**

The Diagnostics register allows to user to check on various internal signals of the ROC chip to verify that it is functioning properly. It may also provide a means to diagnose problems. After power up all bits are zero but they may turn on at any time. In the current version of the ROC, the individual bits have the following meaning:

Bit	Signal	Default	Description
0	Active	0	a TDC or an internal register is active (i.e. it has the token and is sending data)
1	DataReady	0	a TDC or an internal register has set the Data Ready signal
2	TokenAway	0	the ROC has sent out the token to the TDCs
3	TdcAct	0	the OR of the TDC Active signals
4	TdcErr	0	the OR of the TDC Error signals
5	GetData	0	the ROC is asking for data from a TDC or an internal register
6	SerBusy	0	the data serializer is busy sending a word via the FELink (Data/Strobe output lines)
7	SendEvent	0	the ROC received a trigger and is sending out data belonging to this event

Table 19: Read Out Control, Diagnostics register

This register is primarily intended to be used by experts only. The register is read only and does not require to be initialized via the JTAG port. The signals that are attached to the individual bits, may vary between different versions of the ROC (see also diagnostics connector).

### DAC Register (IR=0101)

The DAC register drives the Digital-Analog-Converter chip on the Datimizer card (this is a front-end board for a wire chamber). This register has no use on the CPC card, but it is kept here for compatibility.

Bits	Function	Default	Description
7...0	DAC	0000 0000	Threshold value for discriminators

### Mode Register (IR=0110)

The DAC register drives the Digital-Analog-Converter chip on the Datimizer card (this is a front-end board for a wire chamber). This register has no use on the CPC card, but it is kept here for compatibility.

Bit	Signal	Default	Description
0	ART0	0	Artificial Reset/Trigger selector
1	ART1	0	"
2	ART2	0	"
3	TS3	0	test output on connector roc_ts pin 3
4	TS4	0	test input on connector roc_ts pin 4
5	TS5	0	test input on connector roc_ts pin 5
6	Spare6	0	Spare
7	Spare7	0	Spare

Table 20: Read Out Control, Mode register

The encoding of the ART (artificial reset and trigger) bits is as follows:

ART2	ART1	ART0	Description
0	0	0	no operation
0	0	1	Trigger
0	1	0	bunch count reset: resets only the TDCs
0	1	1	event count reset: reset ROC and the TDCs
1	0	0	global reset: resets ROC and the TDCs
1	1	1	emulate power up reset inside ROC
Others			no operation

### RAND Register (IR=0111)

The RAND register provides a way to set a seed for the internal 9 bit random number generator. The generated number is used as a word count for the fake TDC word unit. When the EnFake bit in the Setup register is set, a random number of fake TDC words is inserted

into the data stream that is sent out via the FELink. The fake TDC word is formatted as "33xxxxxx", where the lower 24 bits are filled with a rotating bit pattern.

Bits	Function	Default	Description
7...0	RAND	0000 0000	Seed for random number generator

Bit 7 and 6 of the seed are AND-ed with the generated random number to form the fake word count for one event, in order to limit the maximum word count. So the maximum word count will either be 63, 127, 191 or 255. Each trigger will get the next number from the pseudo random number generator.

### Bypass Register (IR=1111 or IR='others')

When loading a different pattern than those described above into the instruction register, the function is *Bypass*. This is also mandatory for IR=1111. The bypass register is one bit long and its data is always "0".

### Diagnostics connector

There is a diagnostics connector on the CPC card where one can monitor several signals of the ROC chip.

pin	Signal	Diag bit	Description
1	Vcc	-	Vcc (+ 5 Volt).
2		-	not connected
3	ClkA	-	the clock signal of the ROC
4	Trigger	-	the trigger signal of the ROC and TDCs
5	GetDataIn	-	the internal GetData signal (for bog/eog)
6	DataReady	-	the internal DataReady signal ( " )
7	CTrigger	-	the internal Trigger signal
8	CTokenOut	-	the token signal sent out to the TDCs
9	CTokenIn	-	the token signal coming from the TDCs
10	ROReset	-	the internal ROC Reset signal
11	PowerUp	-	the internal PowerUp signal (active low)
12	SEvent	7	the ROC received a trigger and is sending out data belonging to this event
13	SerBusy	6	the data serializer is busy sending one word via the FELink (Data/Strobe output lines)
14	GetData	5	the ROC is asking for data from a TDC or an internal register
15	TdcErr	4	the OR of the TDC Error signals
16	TdcAct	3	the OR of the TDC Active signals
17	TokenAway	2	the ROC has sent out the token to the TDCs
18	DataReady	1	a TDC or an internal register has set the Data Ready signal
19	Active	0	a TDC or an internal register is active (i.e. it has the token and is preparing data for readout)
20	Gnd	-	Ground (0 Volt)

### TDC programming

The *Ajtag* or *Settings* JTAG port connects to the TDC chips according the order shown in Figure 8. See Figure 5 for the location in the total chain. It is used to load the settings of the TDC chips and check their status. The TDC's JTAG ports are in accordance with the IEEE 1149.1 standard.

The text and tables below are copied from Christiansen's TDC description [<sup>3</sup>] version 0.4, chapter 14. For the most recent information one should use the TDC's manual.

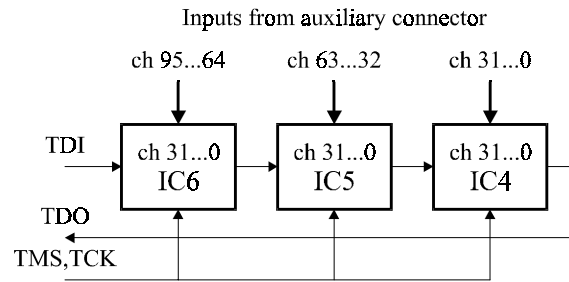


Figure 8: JTAG, the TDC chain

### JTAG Test and Programming Port.

A JTAG (Joint Test Action Group, IEEE 1149.1 standard [3,4]) port is used to program the programmable features in the TDC, and also to get access to test facilities built into the TDC. The JTAG protocol is today an accepted standard for chip and module testing, and easy to use debugging tools are available on PCs. Full boundary scan is supported to be capable of performing extensive testing of TDC modules while located in the system. Testing the functionality of the chip itself is also supported by the JTAG INTEST capability. In addition special JTAG registers have been included in the data path of the chip to be capable of performing effective testing of all the timing registers and embedded memory structures.

Programming of the device is separated into two scan path groups. The setup group consists of setups which can not be changed while the system is actively running (trigger window, looking back window, enable of trigger matching, etc.) and a control group which can be changed during a run (enable and disable of noisy channels). To save silicon area for the large shift registers for the programming data they are not made with two sets of JTAG registers (shift register and update register). This has the inconvenience that if the programming data are read via JTAG (only used during functional testing of devices) one of the programming bits will be lost and the bit sequence will be shifted by one.

An additional scan path is available to read status information from the TDC while it is running (error flags).

<sup>3</sup> A 32 channel general purpose Time to Digital Converter, J. Christiansen CERN/ECP-MIC.

### 14.1. JTAG instructions

The JTAG instruction register is 4 bits long and is in full accordance with the IEEE 1149.1 standard. The JTAG controller macro from ES2 does not support being read from JTAG. When passing through the JTAG state CAPTURE-DR the instruction register gets set to 0001 Bin (IDCODE).

0000	EXTEST.	Boundary scan for test of inter-chip connections on module.
0001	IDCODE.	Scan out of chip identification code.
0010	SAMPLE.	Sample of all chip pins via boundary scan registers.
0011	INTEST.	Using boundary scan registers to test chip itself.
0100...0111	NOT IMPLEMENTED.	
1000	Setup.	Load of setup data.
1001	Control.	Load of control information.
1010	Status.	Read of status information.
1011	Coretest	Access to internal test scan registers.
1100...1110	NOT IMPLEMENTED.	
1111	BYPASS.	

### 14.2. Boundary scan register.

All signal pins of the TDC are passed through JTAG boundary scan registers. All JTAG test modes related to the boundary scan registers are supported (EXTEST, INTEST, SAMPLE).

BSR#	Pin name	Description
0	Enable data_ready	Enable of data_ready driver (not a direct pin).
1	Enable bus	Enable of read-out bus drivers (not a direct pin).
6:2	Vernier[4:0]	Read-out vernier.
22:7	Coarse_out[15:0]	Read-out time coarse out.
28:23	Channel[5:0]	Read-out channel.
29	Rising	Read-out rising edge.
30	Active	TDC active in read-out.
31	Event_end	Read-out event end.
32	Data_ready	Read-out data ready.
33	Error	Programmable error status output.
49:34	Coarse_in[15:0]	Read-out time coarse in (parallel trigger in).
50	Chip_select	Read-out chip select.
51	Next_event	Read-out next event.
52	Get_data	Read-out get data.
53	Trigger	Trigger in
54	Reset	Reset of buffers and counters.
87:55	Hit[32:0]	Hits.
88	Clk	Clock.

### 14.3. ID code

A 32 bit chip identification code can be shifted out when selecting the ID shift chain.

0		Start bit = 1.
11: 1	ES2 code	ES2 manufacturer code = 00001000111.
27: 12	TDC part code	decimal 3500 = binary 0000110110101100
31: 28	Version code	first version = 0000

### 14.4. Setup registers

The JTAG setup scan path is used to load programming data that can not be changed while the TDC is actively running. For testing purposes it is possible to read the status of internal counters via this scan path.

15:0	Yes/no trigger time tag offset[15:0]. (read back of trigger time tag counter)
16	Enable parallel trigger. *1
17	Enable sync trigger. *1
18	Enable serial trigger. *1
34:19	Trigger matching window [15:0]. (read back of active trigger)
35	Enable read-out of start measurements.
36	Enable subtraction of trigger time tag.
37	Enable overlapping triggers.
45:38	Looking back window[7:0]
53:46	Looking ahead window [7:0]
54	Enable subtraction of start time measurement (channel 32)
55	Enable matching
56	Enable automatic reject
72:57	Reject offset[15:0]. (read back of reject counter)
78:73	Adjust_channel[5:0]
86:79	Adjust[7:0]
87	Enable_individual_adjust
89:88	Test mode: 00 = normal mode, 01 = Coarse, Vernier, channel, falling from internal test scan path, 10 = Vernier from internal test scan path, 11 = Toggle between internal test scan path and its inverted value.
94:90	DLL_current_level_b[4:0], charge pump current levels inverted. For normal operation = 11110 bin (minimum current levels).
95	DLL reset
96	Detect falling edge start
97	Detect falling edge odd channels
98	Detect falling edge even channels
99	Detect both edges (all channels)
100	Enable empty start
101	Must be equal to setup bit 54

102	Enable use of double synchronisers.
103	Enable double hit priority queue.
104	Enable start gating
120:105	Coarse count offset[15:0] (read back of coarse count)
121	Enable token read-out mode.
122	Not_locked error mask.
123	Hit_error error mask.
124	Event_buffer_overflow error mask.
125	Trigger_buffer_overflow error mask.
126	Serial_trigger_error error mask.

\*1: Only one of the three trigger modes can be enabled.

#### 14.4.1. Programming of channel adjustment constants.

The programming of the channel adjustment constants are a bit more complicated than the other programmable parameters loaded via the setup scan path. The required 32 eight bit channel adjustment constants are not directly accessible as individual fields of the setup scan path. The 32 channel specific constants are stored in a memory written into via the setup scan path. The adjustment constant for the start channel (or the common adjustment constant) is taken as the last adjustment constant shifted into the setup scan chain (not contained in channel adjustment memory). This approach saves significant silicon area but requires the adjustment constants to be loaded one by one using the serial JTAG protocol. In case only a common adjustment is required the channel adjustment memory can be left non initialized.

Loading of one of the channel adjustment constants into the channel adjustment memory is performed as follows. The channel identifier Adjust\_channel[5:0] and its corresponding adjustment constant Adjust[7:0] must be loaded into the setup scan path. When the setup data have been loaded, and the JTAG controller gives a update signal (by passing through the JTAG state: UPDATE) to the JTAG registers, the adjustment constant will be written into the memory position pointed to by the channel identifier. The other bit fields of the setup data can be left undefined during the loading of the channel specific adjustment constants.

The loading of the common (or start) adjustment constant must be performed as the last loading of the setup scan chain. The channel identifier must in this case be set to Adjust\_channel[5:0]=10 0000 bin to signal that this value shall not be written into the adjustment memory, but must remain in the normal JTAG scan register. In addition all the other programming parameters in the setup chain must have all their parameters set as required for the application.

#### 14.4.2. Performing a reset of the DLL.

The Delay Locked Loop is not initialized when the reset pin of the TDC chip is asserted. The lock tracing of the DLL is a rather slow process and is only required to be performed once after power has been applied. The normal reset pin of the TDC can be used at regular intervals to initialize all counters and buffers in the TDC without having to wait for the DLL to acquire lock.

A reset of the DLL must be performed via the JTAG setup scan path after power up. First a load of the setup registers must be performed with the DLL\_reset signal assigned to a logical one and the required charge pump current levels set to their correct value. For this to perform a real reset of the DLL it is required that the JTAG controller issues an update by

being passed through the update state of the JTAG port controller [3,4]. To enable the DLL to start lock tracing from its reset state the DLL\_reset signal must be released again in the same manner. Now the DLL will start lock tracing and when lock has been obtained the not\_locked status bit will be cleared.

#### 14.5. Control registers

The JTAG control scan path is used to enable/disable channels which can be done while the TDC is actively running.

32:0	Channel enable [32:0]
------	-----------------------

#### 14.6. Status registers

The JTAG status scan path is used to get access to the error status of the TDC while it is running (or after a run).

0	Not_locked.
1	Hit_error.
2	Event_buffer_overflow.
3	Trigger_buffer_overflow.
4	Serial_trigger_error.

#### 14.7. Internal test registers

The JTAG internal test scan path is used to perform extended testing of the TDC chip. The internal scan path gives direct access to the interface between the hit registers and first level buffer logic. It is used in connection with the test mode select bits in the setup scan path.

31:0	Vernier in non encoded form [31:0].
47:32	Coarse[15:0]
53:48	Channel[5:0]
54	Falling edge
55	Select difference (read only)
56	Select start (read only)
57	Store start (read only)
58	Store event (read only)